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01	AMERICAN NATIONAL STANDARD	FC-PI-7 Rev 0.00
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03		
04	American National Standard	
05	for Information Technology–	
06		
07		
08	Fibre Channel –	
09	Physical Interface-7 (FC-PI-7)	
10		
11	1 Scope	
12		
13	This international standard describes the physical interface portions of high performance electrical	
14	and optical link variants that support the higher level Fibre Channel protocols including FC-FS-4 (ref-	
15	erence [23]) and FC-FS-5 (reference [25]).	
16		
17	FC-PI-7 includes 64GFC and 256GFC. 32GFC and 128GFC are described in FC-PI-6 reference [1]	
18	and FC-PI-6P reference [2]) respectively. 16GFC, 8GFC and 4GFC are described in FC-PI-5 (refer-	
19	ence [3]). Older technologies of 2GFC and 1GFC are listed in FC-PI-2 (reference [5]).	
20		
21		
22		
23	2 Normative references	
24		
25	2.1 General	
26		
27	The following standards contain provisions that, through reference in this text, constitute provisions of	
28	this standard. At the time of publication, the editions indicated were valid. Standards are subject to re-	
29	vision, and parties to agreements based on this Standard are encouraged to investigate the possibil-	
30	ity of applying the most recent editions of the following list of standards. Members of IEC and ISO	
31	maintain registers of currently valid International Standards.	
32		
33	Copies of the following documents can be obtained from ANSI: Approved ANSI standards, approved	
34	and draft international and regional standards (ISO, IEC), and other approved standards (including	
35	JIS and DIN).	
36		
37	2.2 Normative references	
38		
39	2.2.1 Approved references	
40		
41	[1] ANSI/INCITS 479-2011, FC-PI-6, Fibre Channel Physical Interfaces - 6	
42	[2] ANSI/INCITS 479-2011, FC-PI-6P, Fibre Channel Physical Interfaces - 6P	
43	[3] ANSI/INCITS 479-2011, FC-PI-5, Fibre Channel Physical Interfaces - 5	
44	[4] ANSI/INCITS 460-2011, FC-PI-3, Fibre Channel Physical Interfaces - 3	
45	[5] ANSI/INCITS 404-2006, FC-PI-2, Fibre Channel Physical Interfaces - 2	
46	[6] ANSI/INCITS TR-35-2006, FC-MJSQ, Fibre Channel Methodologies for Jitter and Signal	
47	Quality Specification	
48	[7] ANSI/INCITS TR-46-2011, FC-MSQS, Fibre Channel Methodologies for Signal Quality	
49	Specification	
50	[8] IEC 60793-1-43, Optical fibers - Part 1-43: Measurement methods and test procedures -	
51	Numerical aperture	
52		
53		

00	[9]	IEC 60793-2-10, Optical fibers - Part 2-10: Product specifications - Sectional specification for category A1 multimode fibers	00
01			01
02	[10]	IEC 60793-2-50, Optical fibers - Part 2-50: Product specifications - Sectional specification for class B single-mode fibers	02
03			03
04			04
05	[11]	IEC 60825-1, Safety of laser products - Part 1: Equipment classification and requirements, latest edition.	05
06			06
07	[12]	IEC 60825-2, Safety of laser products - Part 2: Safety of optical fiber communication systems, latest edition.	07
08			08
09			09
10	[13]	IEC 61280-1-1, Transmitter Output Power Coupled into Single-Mode Fiber Optical Cable	10
11			11
12	[14]	IEC 61280-1-3, Fiber optic communication subsystem basic test procedures - Part 1-3: Test procedures for general communication subsystems - Central wavelength and spectral width measurement.	12
13			13
14			14
15	[15]	IEC 61280-2-2, Fiber optic communication subsystem test procedure - Part 2-2: Digital systems - Optical eye pattern, waveform, and extinction ratio measurements	15
16			16
17	[16]	IEEE Std 802.3™-2012, IEEE Standard for Ethernet.	17
18			18
19	[17]	OIF-CEI-03.0, Common electrical I/O (CEI) - Electrical and jitter interoperability agreements for 6G+ bps, 11G+ bps and 25G+ bps I/O	19
20			20
21	[18]	OIF2010.404.15 OIF CEI-28G-VSR Very Short Reach Interface	21
22			22
23	[19]	TIA-492AAAC, Detail Specification for 850-nm Laser-Optimized, 50-μm core diameter/125-μm cladding diameter class Ia graded-index multimode optical fibers	23
24			24
25	[20]	TIA-492AAAD, Detail Specification for 850-nm Laser-Optimized, 50-μm core diameter/125-μm cladding diameter class Ia graded-index multimode optical fibers suitable for manufacturing OM4 cabled optical fiber	25
26			26
27			27
28	[21]	IEEE P802.3bj, 100 Gb/s Backplane and Copper Cable	28
29	[22]	OIF2014.230.03 OIF-CEI-56G-VSR-PAM4 Very Short Reach Interface	29
30			30
31	[23]	ANSI/INCITS 1861D, FC-FS-4, Fibre Channel Framing and Signaling 4	31
32			32
33	[24]	ANSI/INCITS 1734DT, FC-MSQS-2, Fibre Channel Methodologies for Signal Quality Specification 2	33
34			34
35			35
36	2.3	References under development	36
37			37
38		At the time of publication, the following referenced standards were still under development. For information on the current status of the documents, or regarding availability, contact the relevant standards body or other organization as indicated.	38
39			39
40			40
41	[25]	ANSI/INCITS 1861D, FC-FS-5, Fibre Channel Framing and Signaling 5	41
42			42
43	[26]	IEEE 802.3bs 200 Gb/s and 400 Gb/s Ethernet Task Force	43
44	[27]	ANSI/INCITS _____, FC-MSQS-3, Fibre Channel Methodologies for Signal Quality Specification 3	44
45			45
46			46
47			47
48			48
49			49
50			50
51			51
52			52
53			53

00	3 Definitions and conventions	00
01		01
02	For the purposes of this Standard, the following definitions, conventions, abbreviations, acronyms,	02
03	and symbols apply.	03
04		04
05	3.1 Definitions	05
06	3.1.1 α_T, α_R : alpha T, alpha R; reference points used for establishing signal budgets at the chip	06
07	pins of the transmitter and receiver in an FC device or retiming element.	07
08		08
09	3.1.2 β_T, β_R: beta T, beta R; interoperability points used for establishing signal budget at the disk-	09
10	drive connector nearest the alpha point unless the point also satisfies the definition for delta-	10
11	or gamma when it is either a delta or a gamma point. The beta point specifications are intra-	11
12	enclosure specifications.	12
13		13
14	3.1.3 γ_T, γ_R : gamma T, gamma R; interoperability points used for establishing signal budgets at the	14
15	external enclosure connector.	15
16	3.1.4 δ_T, δ_R : delta T, delta R; interoperability points used for establishing signal budget at the	16
17	internal connector of a removable PMD element.	17
18		18
19	3.1.5 ϵ_T, ϵ_R: epsilon T, epsilon R; interoperability points used for establishing signal budget at-	19
20	internal connectors mainly in blade applications. The epsilon point specifications are for intra-	20
21	enclosure specifications.	21
22		22
23	3.1.6 alpha T, alpha R: see α_T, α_R .	23
24	3.1.7 attenuation: the transmission medium power or amplitude loss expressed in units of dB.	24
25		25
26	3.1.8 average power: the optical power measured using an average-reading power meter when	26
27	transmitting valid transmission characters.	27
28	3.1.9 bandwidth: the difference between the upper -3 dB frequency and the lower -3 dB frequency	28
29	of the amplitude response of a Fibre Channel component.	29
30		30
31	3.1.10 baud: a unit of signaling speed, expressed as the maximum number of times per second the	31
32	signal may change the state of the transmission line or other medium. (Units of baud are	32
33	symbols/sec.) NOTE: With the Fibre Channel transmission scheme, a symbol represents a-	33
34	single transmission bit.	34
35	3.1.11 beta T, beta R: see β_T, β_R.	35
36		36
37	3.1.12 bit error ratio (BER): the probability of a correct transmitted bit being erroneously received	37
38	in a communication system. For purposes of this standard BER is the number of bits output	38
39	from a receiver that differ from the correct transmitted bits, divided by the number of	39
40	transmitted bits.	40
41	3.1.13 bit synchronization: the condition that a receiver is delivering retimed serial data at the	41
42	required BER.	42
43		43
44	3.1.14 byte: an eight-bit entity prior to encoding, or after decoding, with its least significant bit	44
45	denoted as bit 0 and most significant bit as bit 7. The most significant bit is shown on the left	45
46	side unless specifically indicated otherwise.	46
47	3.1.15 bulkhead: the boundary between the shielded system enclosure (where EMC compliance is-	47
48	maintained) and the external interconnect.	48
49		49
50	3.1.16 cable plant: all passive communications elements (e.g., optical fiber, twisted pair, coaxial	50
51	cable, connectors, splices, etc.) between a transmitter and a receiver.	51
52		52
53		53

00	3.1.17 center wavelength (laser): the value of the central wavelength of the operating, modulated	00
01	laser. This is the wavelength where the effective optical power resides. See IEC 61280-1-3	01
02	(reference [14]).	02
03		03
04	3.1.18 character: a defined set of n contiguous bits where n is determined by the encoding	04
05	scheme.	05
06	3.1.19 coaxial cable: an unbalanced electrical transmission medium consisting of concentric	06
07	conductors separated by a dielectric material with the spacings and material arranged to give	07
08	a specified electrical impedance.	08
09		09
10	3.1.20 component: entities that make up the link. Examples are connectors, cable assemblies,	10
11	transceivers, port bypass circuits and hubs.	11
12	3.1.21 connector: electro-mechanical or opto-mechanical components consisting of a receptacle	12
13	and a plug that provides a separable interface between two transmission media segments.	13
14	Connectors may introduce physical disturbances to the transmission path due to impedance	14
15	mismatch, crosstalk, etc. These disturbances may introduce jitter under certain conditions.	15
16		16
17	3.1.22 cumulative distribution function (CDF): the integral of the probability distribution function	17
18	(PDF) from minus infinity to a specific time or from a specific time to plus infinity.	18
19	3.1.23 data dependent pulse width shrinkage (DDPWS): the difference between nominal bit	19
20	period and the minimum value of the zero-crossing-time differences of all adjacent edges in	20
21	an averaged waveform of a repeating data sequence.	21
22		22
23	3.1.24 delta T, delta R: see δ_T , δ_R .	23
24	3.1.25 deterministic jitter: see jitter, deterministic.	24
25		25
26	3.1.26 device: see FC device.	26
27	3.1.27 disparity: the difference between the number of ones and zeros in a Transmission	27
28	Character. See FC-FS-4 (reference []).	28
29		29
30	3.1.28 dispersion: (1) a term in this document used to denote pulse broadening and distortion from	30
31	all causes. The two causes of dispersion in optical transmissions are modal dispersion, due	31
32	to the difference in the propagation velocity of the propagation modes in a multimode fiber,	32
33	and chromatic dispersion, due to the difference in propagation of the various spectral	33
34	components of the optical source. Similar effects exist in electrical transmission lines. (2)	34
35	Frequency dispersion caused by a dependence of propagation velocity on frequency, that	35
36	leads to a pulse widening in a system with infinitely wide bandwidth. The term 'dispersion'	36
37	when used without qualifiers is definition (1) in this document.	37
38	3.1.29 duty cycle distortion (DCD): (1) the absolute value of one half the difference in the average	38
39	pulse width of a '1' pulse or a '0' pulse and the ideal bit time in a clock like (repeating	39
40	0,1,0,1,...) bit sequence. (2) One half of the difference of the average width of a one and the	40
41	average width of a zero in a waveform eye pattern measurement. Definition (2) contains the	41
42	sign of the difference and is useful in the presence of actual data. DCD from definition (2)	42
43	may be used with arbitrary data. DCD is not a level 1 quantity. DCD is considered to be	43
44	correlated to the data pattern because it is synchronous with the bit edges. Mechanisms that	44
45	produce DCD are not expected to change significantly with different data patterns. The	45
46	observation of DCD may change with changes in the data pattern. DCD is part of the DJ	46
47	distribution and is measured at the average value of the waveform.	47
48		48
49	3.1.30 effective DJ: DJ used for level 1 compliance testing, and determined by curve fitting a	49
50	measured CDF to a cumulative or integrated dual Dirac function, where each Dirac impulse,	50
51	located at +DJ/2 and -DJ/2, is convolved with separate half magnitude Gaussian functions	51
52	with standard deviations sigma1 and sigma2. Equivalent to level 1 DJ.	52
53		53

00	3.1.31	enclosure: the outermost electromagnetic boundary (that acts as an EMI barrier) containing one or more FC devices.	00
01			01
02	3.1.32	epsilon T, epsilon R: see ϵ_T, ϵ_R.	02
03			03
04	3.1.33	external connector: a bulkhead connector, whose purpose is to carry the FC signals into and out of an enclosure, that exits the enclosure with only minor compromise to the shield effectiveness of the enclosure.	04
05			05
06			06
07			07
08	3.1.34	extinction ratio: the ratio of the high optical power to the low optical power. See IEC 61280-2-2 (reference [15]).	08
09			09
10	3.1.35	FC-0 level: The level in the Fibre Channel architecture and standards that defines transmission media, transmitters and receivers, and their interfaces. See FC-FS-4 (reference []).	10
11			11
12			12
13	3.1.36	FC-1 level: The level in the Fibre Channel architecture and standards that defines the transmission protocol that includes the serial encoding, decoding, and error control. See FC-FS-4 (reference []).	13
14			14
15			15
16			16
17	3.1.37	FC device: an entity that contains the FC protocol functions and that has one or more of the connectors defined in this document. Examples are: host bus adapters, disk drives, and switches. Devices may have internal connectors or bulkhead connectors.	17
18			18
19			19
20	3.1.38	FC device connector: a connector defined in this document that carries the FC serial data signals into and out of the FC device.	20
21			21
22			22
23	3.1.39	fiber optic cable: a jacketed optical fiber or fibers.	23
24			24
25	3.1.40	gamma T, gamma R: see γ_T , γ_R .	25
26	3.1.41	Golden PLL: this function extracts the jitter timing reference from the data stream under test to be used as the timing reference for the instrument used for measuring the jitter in the signal under test. It conforms to the requirements in 6.10.2 of FC-MJSQ (reference [6]), as modified for 32GFC. For 16GFC and lower speeds the 3dB bandwidth is (nominal signalling rate)/1667. For 32GFC the 3dB bandwidth is (nominal signalling rate)/2805.	26
27			27
28			28
29			29
30			30
31	3.1.42	insertion loss: the ratio (expressed in dB) of incident power at one port to transmitted power at a different port, when a component or assembly with defined ports is introduced into a link or system. May refer to optical power or to electrical power in a specified frequency range. Note the dB magnitude of S12 or S21 is the negative of insertion loss in dB.	31
32			32
33			33
34			34
35			35
36	3.1.43	integrated crosstalk noise: an estimate of the noise due to crosstalk. It is calculated from the S parameters of the channel and takes into account the spectrum, risetime, and amplitude of the crosstalk sources. See clause 10.4 of FC-MSQS (reference [7]).	36
37			37
38			38
39			39
40	3.1.44	insertion loss deviation: the insertion loss deviation ILD is the difference between the measured insertion IL and the fitted insertion loss IL _{fitted} . See clause 10.2.6.4 and clause 12.2 in OIF-CEI-03.0 (reference [17]).	40
41			41
42			42
43	3.1.45	interface connector: an optical or electrical connector that connects the media to the Fibre Channel transmitter or receiver. The connector set consists of a receptacle and a plug.	43
44			44
45			45
46	3.1.46	internal connector: a connector whose purpose is to carry the FC signals within an enclosure (may be shielded or unshielded).	46
47			47
48	3.1.47	internal FC device: an FC device whose FC device connector is contained within an enclosure.	48
49			49
50			50
51	3.1.48	interoperability point: points in a link or TxRx connection for which this standard defines signal requirements to enable interoperability. This includes both compliance points and reference points. See α_T , α_R , β_T , β_R , γ_T , γ_R , δ_T , δ_R , ϵ_T and ϵ_R .	51
52			52
53			53

- 00 **3.1.49 intersymbol interference (ISI):** reduction in the distinction of a pulse caused by overlapping 00
 01 energy from neighboring pulses. (Neighboring means close enough to have significant 01
 02 energy overlapping and does not imply or exclude adjacent pulses—many bit times may 02
 03 separate the pulses especially in the case of reflections). ISI may result in DDJ and vertical 03
 04 eye closure. Important mechanisms that produce ISI are dispersion, reflections, and circuits 04
 05 that lead to baseline wander. 05
- 06 **3.1.50 jitter:** the instantaneous deviations of a signal edge times at a defined signal level of the 06
 07 signal from the reference times. The reference time is the jitter timing reference specified in 07
 08 6.2.3 of FC MJSQ (reference [6]) that occurs under a specific set of conditions. In this 08
 09 document, jitter is defined at the average signal level. 09
 10
- 11 **3.1.51 jitter, bounded uncorrelated (BUJ):** the part of the deterministic jitter that is not aligned in 11
 12 time to the high probability DDJ and DCD in the data stream being measured. Sources of 12
 13 BUJ include, (1) power supply noise that affects the launched signal, (2) crosstalk that occurs 13
 14 during transmission and (3) clipped Gaussian distributions caused by properties of active 14
 15 circuits. BUJ usually is high population DJ, with the possible exception of power supply 15
 16 noise. 16
- 17 **3.1.52 jitter, data dependent (DDJ):** jitter that is added when the transmission pattern is changed 17
 18 from a clock like to a non clock like pattern. For example, data dependent deterministic jitter 18
 19 may be caused by the time differences required for the signal to arrive at the receiver 19
 20 threshold when starting from different places in bit sequences (symbols). DDJ is expected 20
 21 whenever any bit sequence has frequency components that are propagated at different 21
 22 rates. When different run lengths are mixed in the same transmission the different bit 22
 23 sequences (symbols) therefore interfere with each other. Data dependent jitter may also be 23
 24 caused by reflections, ground bounce, transfer functions of coupling circuits and other 24
 25 mechanisms. 25
 26
- 27 **3.1.53 jitter, deterministic (DJ):** jitter with non Gaussian probability density function. Deterministic 27
 28 jitter is always bounded in amplitude and has specific causes. Deterministic jitter comprises 28
 29 (1) correlated DJ (data dependent (DDJ) and duty cycle distortion (DCD)), and (2) DJ that is 29
 30 uncorrelated to the data and bounded in amplitude (BUJ). Level 1 DJ is defined by an 30
 31 assumed CDF form and may be used for compliance testing. See FC MJSQ (reference [6]). 31
 32
- 33 **3.1.54 jitter distribution:** a general term describing either PDF or CDF properties. 33
- 34 **3.1.55 jitter frequency:** the frequency associated with the jitter waveform produced by plotting the 34
 35 jitter for each signal edge against bit time in a continuously running bit stream. 35
- 36 **3.1.56 jitter, non-compensable data dependent, NC-DDJ:** non-compensable data dependent 36
 37 jitter is a measure of any data dependent jitter that is present after processing by the 37
 38 reference receiver. 38
 39
- 40 **3.1.57 jitter, even-odd:** Even-odd jitter is defined as the magnitude of the difference between the 40
 41 average deviation of all even-numbered transitions and the average deviation of all odd- 41
 42 numbered transitions, where determining if a transition is even or odd is based on possible 42
 43 transitions but only actual transitions are measured and averaged. 43
- 44 **3.1.58 jitter, random, RJ:** jitter that is characterized by a Gaussian distribution and is unbounded. 44
 45
- 46 **3.1.59 jitter, sinusoidal (SJ):** single tone jitter applied during signal tolerance testing. 46
 47
- 48 **3.1.60 jitter timing reference:** the signal used as the basis for calculating the jitter in the signal 48
 49 under test. The jitter timing reference has specific requirements on its ability to track and 49
 50 respond to changes in the signal under test. The jitter timing reference may be different from 50
 51 other timing references available in the system. 51
- 52 **3.1.61 jitter tolerance:** the ability of the link or receiver downstream from the receive 52
 53 interoperability point (γ_R , β_R , or δ_R) to recover transmitted bits in an incoming bit stream in the 53

00	presence of specified jitter in the signal. Jitter tolerance is defined by the amount of jitter	00
01	required to produce a specified bit error ratio. The required jitter tolerance performance	01
02	depends on the frequency content of the jitter. Since detection of bit errors is required to	02
03	determine the jitter tolerance, receivers embedded in an FC Port require that the Port be	03
04	capable of reporting bit errors. For receivers that are not embedded in an FC Port the bit	04
05	error detection and reporting may be accomplished by instrumentation attached to the output	05
06	of the receiver. Jitter tolerance is defined at the minimum allowed signal amplitude unless	06
07	otherwise specified. See also signal tolerance.	07
08		08
09	3.1.62 jitter tracking: the ability of a receiver to tolerate low frequency jitter.	09
10	3.1.63 jitter, uncorrelated, UJ: uncorrelated jitter is a measure of any jitter that is not correlated to	10
11	the data stream. See FC MSQS (reference [7]).	11
12		12
13	3.1.64 level:	13
14	1. A document artifact, e.g. FC-0, used to group related architectural functions. No specific	14
15	correspondence is intended between levels and actual implementations.	15
16	2. In FC-PI-6 context, a specific value of voltage or optical power (e.g., voltage level).	16
17	3. The type of measurement: level 1 is a measurement intended for compliance, level 2 is a	17
18	measurement intended for characterization/diagnosis.	18
19	3.1.65 level 1 DJ: term used in this document for the effective DJ value that is used for DJ	19
20	compliance purposes. See jitter, deterministic.	20
21		21
22	3.1.66 limiting amplifier: an active non linear circuit with amplitude gain that keeps the output	22
23	levels within specified levels.	23
24	3.1.67 link:	24
25	1. Two unidirectional fibers transmitting in opposite directions and their associated	25
26	transmitters and receivers.	26
27	2. A duplex TxRx Connection.	27
28		28
29	3.1.68 MB/s: an abbreviation for megabytes (10⁶) per second.	29
30	3.1.69 media: (1) general term referring to all the elements comprising the interconnect. This	30
31	includes fiber optic cables, optical converters, electrical cables, pc boards, connectors, hubs,	31
32	and port bypass circuits. (2) may be used in a narrow sense to refer to the bulk cable material	32
33	in cable assemblies that are not part of the connectors. Due to the multiplicity of meanings for	33
34	this term its use is not encouraged.	34
35		35
36	3.1.70 mode partition noise: noise in a laser based optical communication system caused by the	36
37	changing distribution of laser energy partitioning itself among the laser modes (or lines) on	37
38	successive pulses in the data stream. The effect is a different center wavelength for the	38
39	successive pulses resulting in arrival time jitter attributable to chromatic dispersion in the	39
40	fiber.	40
41	3.1.71 node: a collection of one or more FC ports controlled by a level above FC-2.	41
42		42
43	3.1.72 numerical aperture: the sine of the radiation or acceptance half angle of an optical fiber,	43
44	multiplied by the refractive index of the material in contact with the exit or entrance face. See	44
45	IEC 60793-1-43 (reference [8]).	45
46	3.1.73 OM2: cabled optical fiber containing 50/125 um multimode fiber with a minimum overfilled	46
47	launch bandwidth of 500 MHz-km at 850 nm and 500 MHz-km at 1300 nm in accordance with	47
48	IEC 60793-2-10 Type A1a.1 fiber. See reference [9].	48
49		49
50	3.1.74 OM3: cabled optical fiber containing 50/125 um laser optimized multimode fiber with a	50
51	minimum overfilled launch bandwidth of 1500 MHz-km at 850nm and 500 MHz-km at 1300	51
52	nm as well as an effective laser launch bandwidth of 2000 MHz-km at 850 nm in accordance	52
53	with IEC 60793-2-10 Type A1a.2 fiber. See reference [9].	53

00	3.1.75 OM4: cabled optical fiber containing 50/125 um laser optimized multimode fiber with a	00
01	minimum overfilled launch bandwidth of 3500 MHz-km at 850 nm and 500 MHz-km at 1300	01
02	nm as well as an effective laser launch bandwidth of 4700 MHz-km at 850 nm in accordance	02
03	with IEC 60793-2-10 Type A1a.3 fiber. See reference [9].	03
04		04
05	3.1.76 optical fiber: any filament or fiber, made of dielectric material, that guides light.	05
06		06
07	3.1.77 optical modulation amplitude (OMA): the difference in optical power between the settled	07
08	and averaged value of a long string of contiguous logic one bits and the settled and averaged	08
09	value of a long string of contiguous logic zero bits. See FC-MSQS (reference [7]).	09
10		10
11	3.1.78 optical receiver sensitivity: the minimum acceptable value of received signal at point	11
12	gamma R to achieve a defined level of BER. For 64GFC, this level is BER < 10 ^{-x} . See also	12
13	the definitions for stressed receiver sensitivity and unstressed receiver sensitivity. See FC-	13
14	MSQS (reference [7]) and FC-MSQS-2 (reference [27]).	14
15		15
16	3.1.79 optical path penalty: a link optical power penalty to account for signal degradation other	16
17	than attenuation.	17
18		18
19	3.1.80 optical return loss (ORL): see return loss.	19
20		20
21	3.1.81 OS1: cabled optical fiber containing dispersion unshifted single-mode fiber in accordance	21
22	with IEC 60793-2-50 Type B1.1 fiber specified at 1.0 dB/1.0 dB at 1310nm/1550nm	22
23	respectively. See reference [10].	23
24		24
25	3.1.82 OS2: cabled optical fiber containing dispersion unshifted, low water peak, single-mode fiber	25
26	in accordance with IEC 60793-2-50 Type B1.3 fiber or bend-insensitive fiber in accordance	26
27	with IEC 60793-2-50 Type B6 fiber specified at 0.4 dB/0.4 dB/0.4 dB at	27
28	1310nm/1383nm/1550nm respectively. See reference [10].	28
29		29
30	3.1.83 P_{alloc}: the effective system power/voltage budget used in TWDP and WDP calculations. See	30
31	FC-MSQS (reference [7]).	31
32		32
33	3.1.84 plug: the cable half of the interface connector that terminates an optical or electrical signal	33
34	transmission cable.	34
35		35
36	3.1.85 Port (or FC Port): a generic reference to a Fibre Channel Port. In this document, the	36
37	components that together form or contain the following: the FC protocol function with	37
38	elasticity buffers to re-time data to a local clock, the SERDES function, the transmit and	38
39	receive network, and the ability to detect and report errors using the FC protocol.	39
40		40
41	3.1.86 receiver (Rx): an electronic component (Rx) that converts an analog serial input signal	41
42	(optical or electrical) to an electrical (retimed or non-retimed) output signal.	42
43		43
44	3.1.87 receiver device: the device containing the circuitry accepting the signal from the TxRx	44
45	Connection.	45
46		46
47	3.1.88 receive network: a receive network consists of all the elements between the interconnect	47
48	connector inclusive of the connector and the deserializer or repeater chip input. This network	48
49	may be as simple as a termination resistor and coupling capacitor or this network may be	49
50	complex including components like photo diodes and trans impedance amplifiers.	50
51		51
52	3.1.89 receptacle: the fixed or stationary half of the interface connector that is part of the	52
53	transmitter or receiver.	53

00	3.1.91 reference points: points in a TxRx Connection that may be described by informative	00
01	specifications. These specifications establish the base values for the interoperability points.	01
02	See α_T and α_R .	02
03		03
04	3.1.92 reflectance: the ratio of reflected power to incident power for given conditions of spectral	04
05	composition, polarization and geometrical distribution. In optics, the reflectance is frequently	05
06	expressed as "reflectance density" or in percent; in communications applications it is	06
07	generally expressed as:	07
08		08
09		09
10		10
11		11
12		12
13		13
14	where	14
15	P_r is the reflected power and P_i is the incident power.	15
16		16
17	3.1.93 reflections: power returned by discontinuities in the physical link.	17
18		18
19	3.1.94 repeater: an active circuit designed to modify the (FC) signals that pass through it by	19
20	changing any or all of the following parameters of that signal: amplitude, slew rate, and edge	20
21	to edge timing. Repeaters have jitter transfer characteristics. Types of repeaters include	21
22	Retimers, Reclockers and amplifiers.	22
23		23
24	3.1.95 retimer (RT): a type of repeater specifically designed to modify data edge timing such that	24
25	the output data edges have a defined timing relation with respect to a bit clock derived from a	25
26	timing reference other than the (FC) data at its input. A retimer shall be capable of inserting	26
27	and removing words from the (FC) data passing through it. In the context of jitter	27
28	methodology, a retimer resets the accumulation of jitter such that the output of a retimer has	28
29	the jitter budget of alpha T.	29
30		30
31	3.1.96 return loss: the ratio (expressed in dB) of incident power to reflected power at the same	31
32	port. May refer to optical power or to electrical power in a specified frequency range. Note the	32
33	dB magnitude of S11 or S22 is the negative of return loss in dB.	33
34		34
35	3.1.97 RIN₁₂OMA: relative Intensity Noise. Laser noise in dB/Hz with 12 dB optical return loss, with	35
36	respect to the optical modulation amplitude.	36
37		37
38	3.1.98 RIN₂₀OMA: relative Intensity Noise. Laser noise in dB/Hz with 20 dB optical return loss, with	38
39	respect to the optical modulation amplitude.	39
40		40
41	3.1.99 run length: number of consecutive identical bits in the transmitted signal, e.g., the pattern	41
42	0011111010 has a run lengths of five (5), one (1), and indeterminate run lengths at either	42
43	end.	43
44		44
45	3.1.100 running disparity: a binary parameter indicating the cumulative disparity (positive or	45
46	negative) of all transmission characters since the most recent of (a) power on, (b) exiting	46
47	diagnostic mode, or (c) start of frame. See FC-FS 4 (reference []).	47
48		48
49	3.1.101 signal: the entire voltage or optical power waveforms within a data pattern during	49
50	transmission.	50
51		51
52	3.1.102 signal level: the instantaneous magnitude of the signal measured in the units appropriate	52
53	for the type of transmission used at the point of the measurement. The most common signal	53
	level unit for electrical transmissions is voltage while for optical signals the signal level or	
	magnitude is usually given in units of power: dBm and microwatts.	
	3.1.103 side-mode suppression ratio: ratio of the power in the dominant spectral mode to the	
	power in the strongest side mode.	

- 00 **3.1.104 signal tolerance:** the ability of the link downstream from the receive interoperability point 00
 01 (γ_R , β_R , δ_R , or ϵ_R) to recover transmitted bits in an incoming data stream in the presence of 01
 02 a specified signal. Signal tolerance is defined at specified signal amplitude(s). Since detection 02
 03 of bit errors is required to determine the signal tolerance, receivers embedded in an FC Port 03
 04 require that the Port be capable of reporting bit errors. For receivers that are not embedded 04
 05 in an FC Port the bit error detection and reporting may be accomplished by instrumentation 05
 06 attached to the output of the receiver. See also jitter tolerance. 06
 07
- 08 ~~3.1.105 **special character:** any Transmission Character considered valid by the Transmission-~~ 08
 09 ~~Code but not equated to a Valid Data Byte. Special Characters are provided by the~~ 09
 10 ~~Transmission Code for use in denoting special functions.~~ 10
- 11 **3.1.106 spectral width (RMS):** the weighted root mean square width of the optical spectrum. See 11
 12 IEC 61280-1-3 (reference [14]). 12
 13
- 14 **3.1.107 stressed receiver sensitivity:** the amplitude of optical modulation in the stressed receiver 14
 15 test given in FC-MSQS-2 (reference [27]). 15
- 16 **3.1.108 stressed receiver vertical eye closure power penalty:** the ratio of the nominal optical 16
 17 modulation amplitude to the vertical eye opening in the stressed receiver test. See FC-MSQS 17
 18 (reference [7]). 18
 19
- 20 **3.1.109 synchronization:** bit synchronization, defined above, and/or Transmission-Word 20
 21 synchronization, defined in FC-FS-4 (reference []). An FC-1 receiver enters the state 21
 22 "Synchronization-Acquired" when it has achieved both kinds of synchronization. 22
- 23 **3.1.110 transceiver:** a transmitter and receiver combined in one package. 23
 24
- 25 ~~3.1.111 **transmission symbol bit:** a symbol of duration one unit interval that represents one or~~ 25
 26 ~~more of two logical values, 0 or 1. For example, for 8b10b encoding, one-tenth of a~~ 26
 27 ~~transmission character.~~ 27
- 28 ~~3.1.112 **transmission character:** any encoded character (valid or invalid) transmitted across a~~ 28
 29 ~~physical interface. Valid transmission characters are specified by the transmission code and~~ 29
 30 ~~include data and special characters.~~ 30
 31
- 32 ~~3.1.113 **transmission code:** a means of encoding data to enhance its transmission characteristics.~~ 32
 33 ~~The transmission code specified by FC-FS-4 (reference []) is byte-oriented, with both valid~~ 33
 34 ~~data bytes and special (control) codes encoded into 10-bit transmission characters.~~ 34
- 35 ~~3.1.114 **transmission word:** a string of four contiguous Transmission Characters occurring on~~ 35
 36 ~~boundaries that are zero modulo 4 from a previously received or transmitted Special~~ 36
 37 ~~Character.~~ 37
 38
- 39 ~~3.1.115 **transmit network:** a transmit network consists of all the elements between a serializer or~~ 39
 40 ~~repeater output and the connector, inclusive of the connector. This network may be as simple~~ 40
 41 ~~as a pull-down resistor and ac capacitor or this network may include laser drivers and lasers.~~ 41
- 42 **3.1.116 transmitter (Tx):** a circuit (Tx) that converts a logic signal to a signal suitable for the 42
 43 communications media (optical or electrical). 43
 44
- 45 **3.1.117 transmitter device:** the device containing the circuitry on the upstream side of a TxRx 45
 46 connection. 46
- 47 **3.1.118 transmitter and dispersion penalty (TDP):** TDP is a measure of the penalty due to a 47
 48 transmitter and its specified worst-case medium, with a standardized reference receiver. See 48
 49 IEEE 802.3, clause 52.9.10. See reference [16]. 49
 50
- 51 ~~3.1.119 **transmitter waveform and dispersion penalty (TWDP):** TWDP is a measure of the~~ 51
 52 ~~deterministic penalty of the waveform from a particular transmitter and reference emulated~~ 52
 53 ~~multimode fibers or metallic media, with a reference receiver.~~ 53

- 00 **3.1.120** ~~**T_{rise} / T_{fall}**: the adjusted 20% to 80% rise and fall time of the optical signal.~~ 00
 01 **3.1.121** ~~**TR_{filter} / TF_{filter}**: the measured 20% to 80% rise or fall time of a fourth-order Bessel- 01
 02 Thomson filter with a step input.~~ 02
 03 03
 04 **3.1.122** ~~**TR_{meas} / TF_{meas}**: the measured 20% to 80% rise or fall time of the optical signal.~~ 04
 05 05
 06 **3.1.123** **TxRx connection**: the complete signal path between a transmitter in one FC device and a 06
 07 receiver in another FC device. 07
 08 **3.1.124** **TxRx connection segment**: that portion of a TxRx connection delimited by separable 08
 09 connectors or changes in media. 09
 10 **3.1.125** **unit interval (UI)**: the nominal duration of a single transmission **bitsymbol**. 10
 11 11
 12 **3.1.126** **unstressed receiver sensitivity**: the amplitude of optical modulation in the unstressed 12
 13 sensitivity receiver test in. See FC-MSQS-2 (reference [27]). 13
 14 **3.1.127** **voltage modulation amplitude (VMA)**: VMA is the difference in electrical voltage between 14
 15 the stable one level and the stable zero level, see FC-MSQS (reference [7]). 15
 16 16
 17 **3.1.128** ~~**waveform distortion penalty (WDP)**: WDP is a measure of the deterministic penalty of a 17
 18 waveform with a reference equalizing receiver.~~ 18
 19 **3.1.129** ~~**word**: in Fibre Channel protocol, a string of four contiguous bytes occurring on boundaries 19
 20 that are zero modulo 4 from a specified reference.~~ 20
 21 21
 22 22

23 3.2 Editorial conventions 23

24 3.2.1 Conventions 24

25 In this Standard, a number of conditions, mechanisms, parameters, states, or similar terms are print- 25
 26 ed with the first letter of each word in upper-case and the rest lower-case (e.g., TxRx connection). 26
 27 Any lower-case uses of these words have the normal technical English meanings. 27
 28 28
 29 29

30 Numbered items in this Standard do not represent any priority. Any priority is explicitly indicated. 30
 31 31

32 In case of any conflict between figure, table, and text, the text takes precedence. Exceptions to this 32
 33 convention are indicated in the appropriate sections. 33

34 In the figures, tables, and text of this document, the most significant bit of a binary quantity is shown 34
 35 on the left side. Exceptions to this convention are indicated in the appropriate sections. 35

36 The ISO convention of numbering is used, i.e. the ten-thousands and higher multiples are separated 36
 37 by a space. A period is used as the decimal demarcation. A comparison of the American and ISO 37
 38 conventions are shown below: 38
 39 39

40 **Table 1 – ISO convention** 40

Alternative ISO	ISO as used in this document	American
2 048	2 048	2048
10 000	10 000	10,000
1 323 462,9	1 323 462.9	1,323,462.9

48 3.2.2 Keywords 48

49 **3.2.2.1** **invalid**: Used to describe an illegal or unsupported bit, byte, word, field or code value. 49
 50 Receipt of an invalid bit, byte, word, field or code value shall be reported as an error. 50
 51 51
 52 52
 53 53

00	3.2.2.2 ignored: Used to describe a bit, byte, word, field or code value that shall not be examined	00
01	by the receiving port. The bit, byte, word, field or code value has no meaning in the specified	01
02	context.	02
03	3.2.2.3 mandatory: A keyword indicating an item that is required to be implemented as defined in	03
04	this standard.	04
05	3.2.2.4 may: A keyword that indicates flexibility of choice with no implied preference (equivalent to	05
06	“may or may not”).	06
07	3.2.2.5 may not: A keyword that indicates flexibility of choice with no implied preference	07
08	(equivalent to “may or may not”).	08
09	3.2.2.6 NA: A keyword indicating that this field is not applicable.	09
10	3.2.2.7 obsolete: A keyword indicating that an item was defined in a prior Fibre Channel standard	10
11	but has been removed from this standard.	11
12	3.2.2.8 optional: Characteristics that are not required by FC-PI-6. However, if any optional	12
13	characteristic is implemented, it shall be implemented as defined in FC-PI-6.	13
14	3.2.2.9 reserved: A keyword referring to bits, bytes, words, fields, pins and code values that are set	14
15	aside for future standardization.	15
16	3.2.2.10 shall: A keyword indicating a mandatory requirement. Designers are required to	16
17	implement all such mandatory requirements to ensure interoperability with other products	17
18	that conform to this standard.	18
19	3.2.2.11 should: A keyword indicating flexibility of choice with a strongly preferred alternative;	19
20	equivalent to the phrase “it is strongly recommended”.	20
21	3.2.2.12 should not: A keyword indicating flexibility of choice with a strongly preferred alternative;	21
22	equivalent to the phrase “it is strongly recommended not to”.	22
23	3.2.2.13 vendor specific: Functions, code values, and bits not defined by this standard and set	23
24	aside for private usage between parties using this standard.	24
25	3.2.3 Abbreviations, acronyms, and symbols	25
26	Abbreviations, acronyms and symbols applicable to this standard are listed in table 2. Definitions of	26
27	several of these items are included in subclause 3.1.	27
28		28
29		29
30		30
31		31
32		32
33		33
34		34
35		35
36		36
37		37
38		38
39		39
40		40
41		41
42		42
43		43
44		44
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46		46
47		47
48		48
49		49
50		50
51		51
52		52
53		53

3.2.3.1 Acronyms and other abbreviations

Table 2 – Acronyms and other abbreviations

Bd	baud
BER	bit error ratio
BUJ	bounded uncorrelated jitter
CDF	cumulative distribution function
dB	decibel
dBm	decibel (relative to 1 mW)
DCD	duty cycle distortion
DDJ	data dependent jitter
DDPWS	data dependent pulse width shrinkage
DJ	deterministic jitter
DUT	device under test
EIA	Electronic Industries Association
EMC	electromagnetic compatibility
EMI	electromagnetic interference
FC	Fibre Channel
FEC	Forward error correction
GBd	gigabaud
hex	hexadecimal notation
IGN	integrated crosstalk noise
ILD	insertion loss deviation
IEEE	Institute of Electrical and Electronics Engineers
ITU-T	International Telecommunication Union - Telecommunication Standardization (formerly CCITT)
JBOD	Just Bunch of Disks
LOS	loss of signal
LW	long wavelength
MB	megabyte = 10 ⁶ bytes
MBd	megabaud
MM	multimode
NA	not applicable
NC-DDJ	non-compensable data dependent jitter
NEXT	near-end crosstalk
OMA	optical modulation amplitude
PMD	physical medium dependent
ppm	parts per million
RFI	radio frequency interference
RIN	relative intensity noise
RJ	random jitter
RMS	root mean square
RN	relative noise
Rx	receiver
SERDES	Serializer/Deserializer
SM	single-mode
S/N(SNR)	signal-to-noise ratio
SW	short wavelength
TCTF	transmitter compliance transfer function
TDP	transmitter and dispersion penalty
TDR	time domain reflectometry
TIA	Telecommunication Industry Association
TJ	total jitter
TWDP	transmitter waveform and distortion penalty
Tx	transmitter
TxRx	a combination of transmitter and receiver
UI	unit interval = 1 bit period
UJ	uncorrelated jitter
ULP	Upper Level Protocol

Table 2 – Acronyms and other abbreviations

VECP	vertical eye closure penalty
WDP	waveform distortion penalty

3.2.3.2 Signaling rate abbreviations

Abbreviations for the signaling rates are frequently used in this document. Table 3 shows the abbreviations that are used and the corresponding signalling rates.

Table 3 – Signaling rate abbreviations

Abbreviation	Signaling rate	Data rate
1GFC	1 062.5 MBd	100 MB/s
2GFC	2 125 MBd	200 MB/s
4GFC	4 250 MBd	400 MB/s
8GFC	8 500 MBd	800 MB/s
16GFC	14 025 MBd	1 600 MB/s
32GFC	28 050 MBd	3 200 MB/s
64GFC	28 900 MBd	6 400 MB/s
128GFC	112 200 MBd	12 800 MB/s
256GFC	115 600 MBd	25 600 MB/s

00 4 FC-PI-7 functional characteristics 00

01 4.1 General characteristics 01

02 Fibre Channel is structured as a set of hierarchical functions as illustrated in Figure 1. The FC-PI-x 02
 03 standards define the physical link, the lowest level denoted FC-0, in the Fibre Channel system. The 03
 04 physical layer interface is designed for flexibility and allows the use of several physical interconnect 04
 05 technologies to meet a wide variety of system application requirements. 05
 06 06

07 The FC-FS-x standards define the signaling protocol and services at the next higher levels. 07

08 Transmission codes and Forward Error Correction (FEC), where applicable, are defined in the FC- 08
 09 FS-x standards. 09
 10 10

11 FC-PI-7 describes the physical link for serial and parallel data streams supporting a signaling rate of 11
 12 64GFC and 256GFC respectively in multimode and single mode fibers as defined in 4.13. Serial lane 12
 13 variants include, 64GFC-SW for MM variant and 64GFC-LW and 64GFC-PSM for single mode vari- 13
 14 ant. Parallel lane variants include, 256GFC-SW4 for MM variant with channels consisting of multiple 14
 15 multimode fibers in a cable, 256GFC-PSM-4 for SM variant with channels consisting of multiple sin- 15
 16 gle mode fibers in a cable, and 256GFC-CWDM4 for SM variant using duplex channels with wave- 16
 17 length division multiplexing. Lane skew restrictions for parallel data stream variants are shown in 4.14 17
 18 18

19 19
 20 Fibre Channel 64GFC and 256GFC links use 256B/257B transmission code; see FC-FS-4 (reference 20
 21 [23]) and FC-FS-5 (reference [25]). This code includes Forward Error Correction (FEC) Reed Solo- 21
 22 mon (544,514) which is required to achieve the link BER objectives. The BER of each TxRx connec- 22
 23 tion in a 64GFC link, as observed prior to error correction, is defined to be 1.09×10^{-4} or better. It is the 23
 24 combined responsibility of the component suppliers and the system integrator to ensure that this level 24
 25 of service is provided at every port in a given Fibre Channel installation. When these conditions are 25
 26 satisfied, it is expected that the link BER after error correction will be not worse than 10^{-15} . 26
 27 27

28 64 GFC has transmitter and receiver clock tolerances of ± 100 ppm. A TxRx Connection bit error ratio 28
 29 (BER) of $\leq 1.09 \times 10^{-4}$ as measured at its receiver is supported. The basis for the BER is the encoded 29
 30 serial data stream on the transmission medium during system operation. 30
 31 31

32 FC-PI-7 defines ten different specific physical locations in the FC system. Eight are interoperability 32
 33 points and two are reference points. No interoperability points are required for closed or integrated 33
 34 links and FC-PI-7 is not required for such applications. For closed or integrated links the system de- 34
 35 signer shall ensure that a BER of better than 1.09×10^{-4} is delivered. 35
 36 36

37 37

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51 51

52 52

53 53

4.2 Compliance test points

The requirements specified in FC-PI-7 shall be satisfied at separable connectors where interoperability and component level interchangeability within the link are expected. A compliance point is a physical position where the specification requirements are met. The compliance points are defined at separable connectors, since these are the points where different components can easily be added, changed, or removed. There is no maximum number of interoperability points between the initiating FC device and the addressed FC device as long as (1) the requirements at the interoperability points are satisfied for the respective type of interoperability point and (2) the end to end signal properties are maintained under the most extreme allowed conditions in the system. The description and physical location of the specified interoperability points are detailed in clause 5.13 of FC-PI-5 (reference [3]). All specifications are at the interoperability points in a fully assembled system as if measured with a non-invasive probe except where otherwise described. Figure 2 shows the reclocker locations for 64GFC multi-mode and single-mode variants.

It is the combined responsibility of the component (the separable hardware containing the connector portion associated with an interoperability point) supplier and the system integrator to ensure that intended interoperability points are identified to the users of the components and system. This is required because not all connectors in a link are interoperability points and similar connectors and connector positions in different applications may not satisfy the FC-PI-7 requirements.

The signal and return loss requirements in this document apply under specified test conditions that simulate some parts of the conditions existing in service. This simulation includes, for example, du-

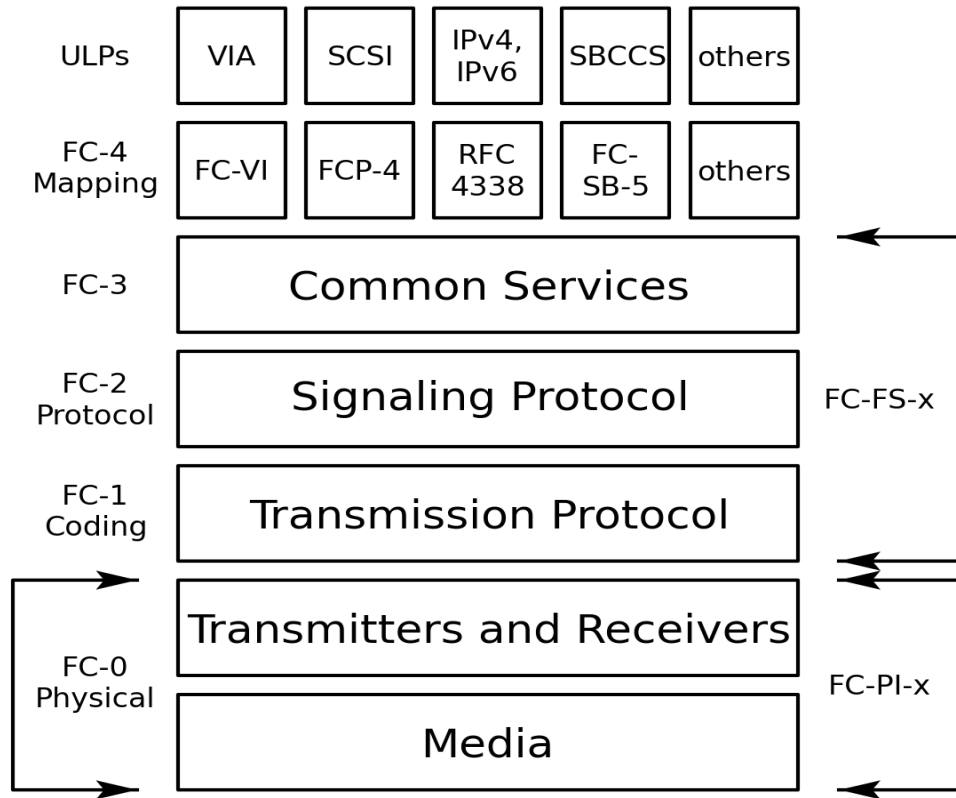


Figure 1 – Fibre Channel hierarchy

plex traffic on all Ports and under all applicable environmental conditions. Effects caused by other features existing in service such as non-ideal return loss in parts of the link that are not present when measuring signals in the specified test conditions are included in the specifications themselves. This methodology is required to give each side of the interoperability point requirements that do not depend on knowing the properties of the other side. In addition, it allows measurements to be performed under conditions that are accessible with practical instruments and that are transportable between measurement sites.

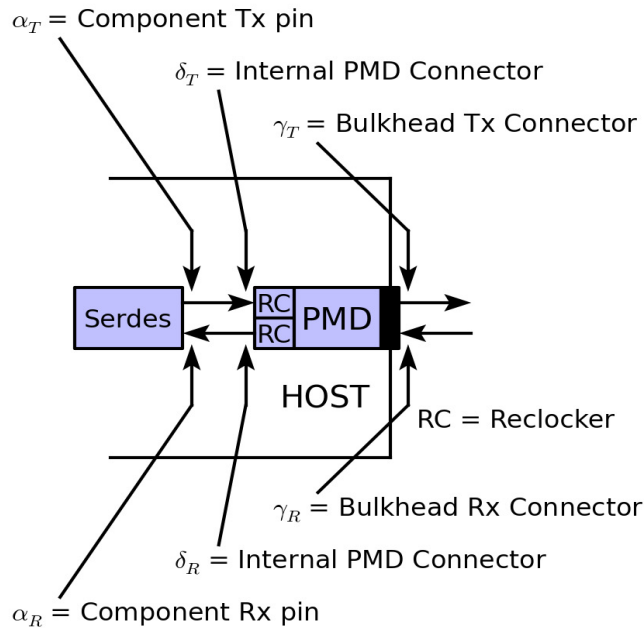


Figure 2 – Reclocker location for all 32GFC PMDs

Measuring signals in an actual functioning system at an interoperability point does not verify compliance for the components on either side of the interoperability point although it does verify that the specific combination of components in the system at the time of the measurement produces compliant signals. Interaction between components on either side of the interoperability point may allow the signal measured to be compliant but this compliance may have resulted because one component is out of specification while the other is better than required.

The interface to FC-FS-4 and FC-FS-5 occur at the logical encoded data interfaces. As these are logical data constructs, no physical implementation is implied by FC-FS-4 and FC-FS-5. FC-PI-7 is written assuming that the same single serial data stream exists throughout the link as viewed from the interoperability points. Other possible schemes for transmitting data, for example using parallel paths, are not defined in FC-PI-7 but could occur at intermediate places between interoperability points.

Physical links have the following general requirements:

- Physical point-to-point data links; no multidrop connections along the serial path.
- Signal requirements shall be met under the most extreme specified conditions of system noise and with the minimum compliant quality signal launched at upstream interoperability points.
- All users are cautioned that detailed specifications shall take into account end-of-life worst case values (e.g., manufacturing, temperature, power supply).

The interface between FC-PI-7 and protocols defined in FC-FS-4 and FC-FS-5 are intentionally structured to be technology and implementation independent. That is, the same set of commands and services may be used for all signal sources and communication schemes applicable to the technology of a particular implementation. As a result of this, all safety or other operational considerations that may be required for a specific communications technology are to be handled by the FC-PI-7 clauses associated with that technology. An example of this would be ensuring that optical power levels associated with eye safety are maintained.

4.3 FC-0 states

4.3.1 Transmitter states

The transmitter device is controlled by the FC-1 level. Its function is to convert the serial data received from the FC-1 level into the proper signal types associated with the transmission media.

4.3.2 Receiver states

The function of the receiver device is to convert the incoming data from the form required by the communications media employed, retime the data, and present the data and an associated clock to the FC-1 level.

4.4 Limitations on invalid code

FC-0 does not detect transmit code violations, invalid ordered sets, or any other alterations of the encoded bit stream. However, it is recognized that individual implementations may wish to transmit such invalid bit streams to provide diagnostic capability at the higher levels. Any transmission violation, such as invalid ordered sets, that follow valid character encoding rules shall be transparent to FC-0. Invalid character encoding could possibly cause a degradation in receiver sensitivity and increased jitter resulting in increased BER or loss of bit synchronization.

4.5 Receiver stabilization time

The time interval required by the receiver from the initial receipt of a valid input to the time that the receiver is synchronized to the bit stream and delivering valid retimed data within the BER requirement, shall not exceed 20 ms. Should the retiming function be implemented in a manner that requires direction from a higher level to start the initialization process, the time interval shall start at the receipt of the initialization request.

4.6 Loss of signal (Rx_LOS) function

The FC-0 may optionally have a loss of signal function. If implemented, this function shall indicate when a signal is absent at the input to the receiver. The activation level shall lie in a range whose upper bound is the minimum specified sensitivity of the receiver and whose lower bound is defined by a complete removal of the input connector. While there is no defined hysteresis for this function there shall be a single transition between output logic states for any monotonic increase or decrease in the input signal power occurring within the reaction time of the signal detect circuitry.

4.7 Speed agile ports that support speed negotiation

This subclause specifies the requirements on speed agile ports that support speed negotiation.

- a) The port transmitter shall be capable of switching from compliant operation at one speed to compliant operation at a new speed within X ms from the time the speed negotiation algorithm asks for a speed change for 16GFC. A repeater shall achieve compliant operation within X ms following an application of a compliant signal at its input. For 16GFC and 32GFC, the transmitter stabilization time shall be X ms or less (allowing up to two repeaters in the path).

- 00 b) The port receiver shall attain Transmission_Word synchronization within the receiver stabiliza- 00
 01 tion time (sub-clause 4.5) when presented with a valid input stream or from the time the algo- 01
 02 rithm asks for a receiver speed change if the input stream is at the new receive rate set by the 02
 03 port implementing the algorithm. 03
 04 c) The port transmitter and port receiver shall be capable of operating at different speeds at the 04
 05 same time during speed negotiation. 05
 06 d) The transmit training signal is used for speed negotiation for 64GFC. The transmit training sig- 06
 07 nal is defined in FC-FS-5 (reference [25]). 07
 08 08
 09 09

10 4.8 Transmission codes 10

11 64GFC and 256GFC variants rely on the implementation of FEC, transcoding, and scrambling as de- 11
 12 fined in FC-FS-5 (reference [25]). The actual FEC, transcoding, and scrambling hardware is at the 12
 13 FC-1 layer and is not defined in FC-PI-7. 13
 14 14
 15 15

16 4.9 Frame scrambling and emission lowering protocol 16

17 64GFC and 256GFC variants use coding and scrambling that is inherent in the code as defined in 17
 18 FC-FS-5 (reference [25]). 18
 19 19
 20 20

21 4.10 Speed negotiation and transmitter training 21

22 For 64GFC and 256GFC variants the transmitter training signal (TTS) shall be used for speed nego- 22
 23 tiation for both optical and electrical links. If the link is a passive electrical link, the transmit training 23
 24 signal will be used for speed negotiation and then transmit training will be done. If the link is an opti- 24
 25 cal link, the transmit training signal is used for speed negotiation and transmit training is not per- 25
 26 formed. The transmit training signal consists of a frame marker, control field, status field, and training 26
 27 pattern. 27
 28 28

29 ~~The frame marker consists of a signal that is 16UI high and 16UI low. The control field and the status~~ 29
 30 ~~field are both 16 bit fields. The control and status field are Differential Manchester Encoded (DME). A~~ 30
 31 ~~DME bit has a length of 8UI and the following properties:~~ 31

- 32 1. ~~There is a data transition at each cell boundary.~~ 32
- 33 2. ~~A mid-cell data transition signals a logic 1.~~ 33
- 34 3. ~~The absence of a mid-cell data transition signals a logic 0.~~ 34
- 35 35

36 ~~The DME encoded status and control field is 256UI.~~ 36

37 ~~The training pattern is 4096UI, 4094UI of PRBS11 followed by 2UI of 0.~~ 37
 38 38

39 ~~During speed negotiation for 64GFC, the previously reserved bits 14, 15 in the control field are set to~~ 39
 40 ~~1 to serve as an extended marker. The speed negotiation bit 14 in the status field is set to 1. The oth-~~ 40
 41 ~~er bits in the control and status field are set to 0.~~ 41

42 ~~The table below highlights the bit sequence for the frame marker, control, and status fields during~~ 42
 43 ~~speed negotiation. This is followed by the 4096UI training pattern. This sequence is repeated until~~ 43
 44 ~~speed negotiation is completed.~~ 44
 45 45
 46 46
 47 47
 48 48
 49 49
 50 50
 51 51
 52 52
 53 53

Table 4 – Transmitter training signal frame marker, control, and status field bit sequence

Frame marker, control status field bits	Identifier
1111 1111 1111 1111	Frame Marker
0000 0000 0000 0000	
1111 0000 1111 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 0000	
1111 1111 0000 1111	Extended Marker
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	Speed negotiation bit high
0000 0000 1111 1111	
0000 0000 1111 1111	
0000 0000 1111 1111	

The table gives an indication of the spectral content of the 288UI that comprise the training frame during speed negotiation. This is then followed by 4096UI training pattern which is composed of PRBS11.

4.11 Forward error correction (FEC)

64GFC and 256GFC variants rely on the implementation of FEC as defined in ~~FC-FS-4 (reference ISO-Text) and~~ FC-FS-5 (reference [25]) The actual FEC hardware is at the FC-1 layer and is not defined in FC-PI-6Z.

PRBS11.

4.12 Test Patterns

64GFC and 256GFC variants shall use the test patterns stated in FC-MSQS-3 (reference [27]).

4.13 FC-PI-7 variants

Table 5 and list variants by FC-PI-7 nomenclature, a reference to the clause containing the detailed requirements, and some key parameters that characterize the variant. The lengths specified in table 5 and are the minimum lengths supported with transmitters, media, and receivers all simultaneously operating under the most degraded conditions allowed. Longer lengths may be achieved by restricting parameters in the transmitter, media, or receiver. If such restrictions are used on the link components then interoperability at interoperability points within the link and component level interchangeability within the link is no longer supported by this standard.

Table 5 – Fibre Channel Variants in FC-PI-7 serial lane

SM OS1, OS2	64GFC-LW 1 300 nm 0.5 m-10 km sub-clause 5.4
	64GFC-PSM 1 300 nm 0.5 m-500 m sub-clause 5.4
MM 50 μm OM3	64GFC-SW 850 nm 0.5 m-70 m sub-clause 5.5
MM 50 μm OM4, OM5	64GFC-SW 850 nm 0.5 m-100 m sub-clause 5.5

Table 6 – Fibre Channel Variants in FC-PI-7 parallel lanes

SM OS1, OS2	256GFC-CWDM4 1 300 nm 0.5 m-2 km sub-clause 5.4
	256GFC-PSM4 1 300 nm 0.5 m-500 m sub-clause 5.4
MM 50 μm OM3	256GFC-SW4 850 nm 0.5 m-70 m sub-clause 5.5
MM 50 μm OM4, OM5	256GFC-SW4 850 nm 0.5 m-100 m sub-clause 5.5

00 4.14 Skew constraints

01 The skew (relative delay) between the lanes must be kept within limits so that the information on the
02 lanes can be reassembled by the RS-FEC sublayer. Skew is defined as the difference in the times of
03 the earliest lane and the latest lane for a one to zero transition. Skew variation may be introduced due
04 to variations in electrical, thermal, or environmental characteristics. Skew variation is defined as the
05 change in skew between any lane and any other lane over the entire time that the link is in opera-
06 tion. Skew and skew variation must be kept within limits as shown in table 7. See Fig. 4 in FC-PI-6P,
07 (reference [2]).
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00 **Table 7 – Skew and skew variation constraints for**
 01 **256GFC**

Test Point	Skew	Skew variation
δ_T	ns	ps
γ_T		
γ_R		
δ_R		
α_R		

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4.15 MPO optical interface

Mechanical, optical performance, and intermatability for the MPO connector system are specified in IEC 61754-7-1 (reference Spec Ref #n). Figure 3 shows the MPO optical interface. The PSM4 optical transceiver requires an MPO female plug connector with a down-angled interface, whereas the multi-mode optical transceiver requires a flat interface.

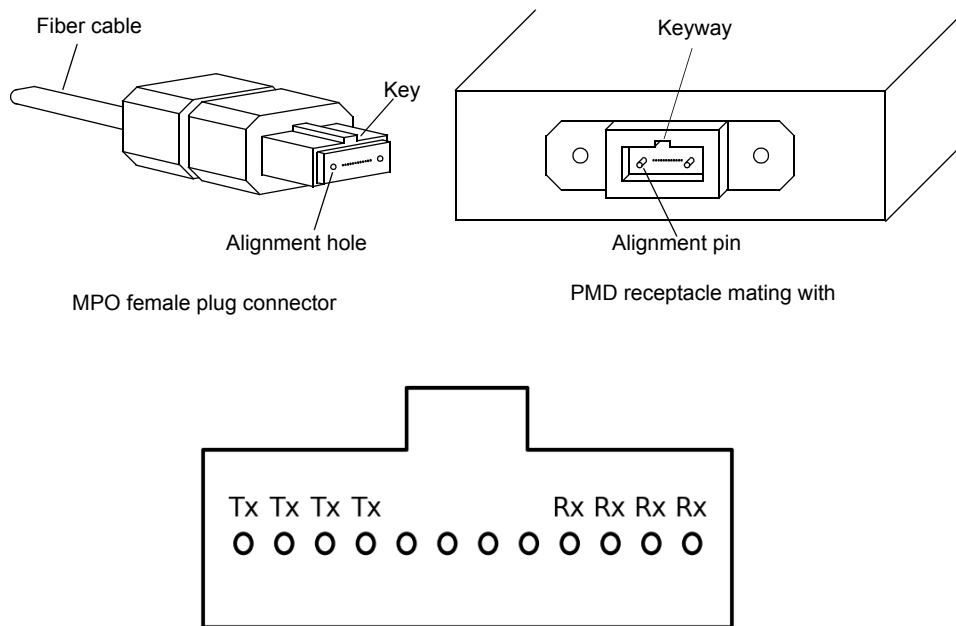


Figure 3 – PMD

00 5 Optical interface specification 00

01 5.1 TxRx connections 01

02 Clause 5 defines the optical signal characteristics at the interface connector. Each conforming optical 02
 03 FC port shall comply with the requirements specified in clause 5 and other applicable clauses. Fibre 03
 04 Channel 64GFC and 256GFC optical links require forward error correction (FEC) to achieve link BER 04
 05 objectives. In the absence of forward error correction, Fibre Channel optical links shall not exceed a 05
 06 BER of 1.09×10^{-4} under any compliant conditions; see FC-MSQS-3 (reference [27]). The parameters 06
 07 specified in this clause support meeting that requirement. 07

08 A link, or TxRx connection, may be divided into TxRx connection segments; see figure 10 in FC-PI-5 08
 09 (reference [1]). In a single TxRx connection individual TxRx connection segments may be formed 09
 10 from differing media and materials, including traces on printed wiring boards and optical fibers. This 10
 11 clause applies only to TxRx connection segments that are formed from optical fiber. 11
 12

13 If electrically conducting TxRx connection segments are required to implement these optical variants, 13
 14 they shall meet the specifications of the appropriate electrical variants defined in clause 6.1.09x10-4 14
 15

16 5.2 Laser safety issues 16

17 Optical transceivers shall conform to Hazard Level 1M laser requirements as defined in IEC 60825-1 17
 18 (reference [11]) and IEC 60825-2 (reference [12]) under any condition of operation. 18
 19

20 5.3 Optical Signal Modulation Format 20

21 A four level pulse amplitude modulation or PAM-4 is the modulation format utilized in all the optical 21
 22 variants defined in PI-7. To generate a PAM-4 signal, two logical bits are mapped to a Gray-coded 22
 23 symbol described in IEEE 802.3bs (reference [26]) and FC-FS-5 (reference [25], IEEE. An non-cor- 23
 24 rupted PAM-4 signal is depicted in Fig. 3 a. The PAM-4 levels 0 and 3 represent the lowest and max- 24
 25 imum optical power. Additionally, the levels 0 and 3 can represent the most negative or most positive 25
 26 voltage when evaluated after the O/E conversion. 26
 27

28 PAM-4 signal generates three eye diagrams: top, middle and bottom, as shown in Fig 3 b. These 28
 29 eyes can present different height and width. Moreover, signals produced by direct modulated lasers 29
 30 can produce eye skew, see FC-MSQS-3 (reference [27]), which penalize the optimum sampling of 30
 31 the signals. 31

32 Transmission dispersion eye closure for PAM-4 or, TDECQ, see IEEE 802.3bs (reference [26]), is 32
 33 used to evaluate the quality of the received optical signal, by estimating the total power budget penal- 33
 34 ty of the received eyes in comparison with an ideal PAM-4 eye. The computation of TDECQ, utilizes 34
 35 measurements average power and outer OMA as well as BER requirements in absence of error cor- 35
 36 rection codes. Note that PI-7 channels, TDECQ is computed using a BER of 1.09×10^{-4} instead of the 36
 37 BER of 2.4×10^{-4} used in reference [26]. For each optical variant, TDECQ as well as the average pow- 37
 38 er and OMA is specified in the tables shown in clause 5. 38
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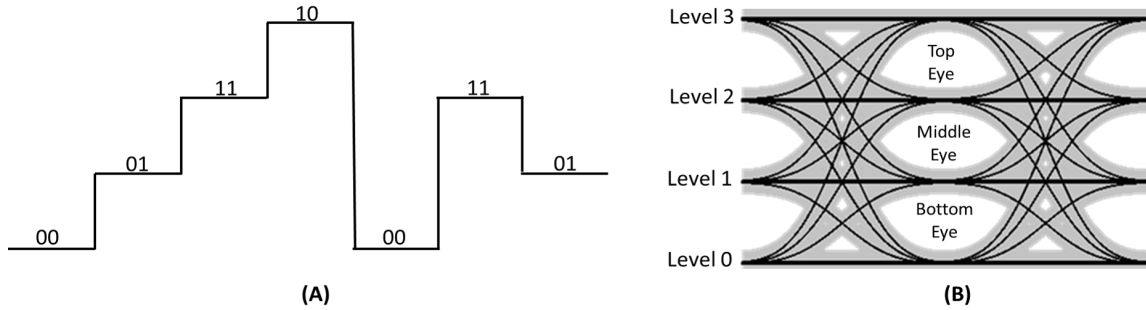


Figure 4 – Optical Eye Diagram of a PAM-4 Signal

5.4 SM data links

5.4.1 SM general information

Table 8 gives the variant name, a general link description, and the gamma compliance point specifications for 10-km single-mode optical fiber links running at 64GFC.

Table 9 gives the variant name, a general link description, and the gamma compliance point (figure 4) specifications for single-mode optical fiber links running at 256GFC-CWDM4.

Table 10 gives the variant name, a general link description, and the gamma compliance point (figure 4) specifications for single-mode optical fiber links running at 256GFC-PSM4.

5.4.2 SM optical output interface

The optical transmit signal is defined at the output end of a patch cord between one half and five meters in length.

5.4.3 SM optical input interface

The receiver shall operate within the uncorrected BER requirement of 1.09×10^{-41} when the input power falls within the range given in table 8 for 64GFC, table 9 for 256GFC-CWDM4 and table 10 for 256GFC-PSM4.

5.4.3.1 64GFC-LW1

Table 8 – Single-mode link classes (OS1, OS2)

Single mode link parameters (note 1)	Unit	64GFC-LW1	Note
Nominal signaling rate	MBd	28 900	2
Operating distance	m	0.5 -10 000	
Transmitter (gamma-T)			
Type	Laser		
Center wavelength, max.	nm	1325	
Center wavelength, min.	nm	1295	
Optical modulation amplitude, min.	mW(dBm)	TBD	
Side-mode suppression ratio, min.	dB	TBD	
-20 dB spectral width, max.	nm	TBD	
Average launched power, max.	dBm	TBD	
Average launched power, min.	dBm	TBD	
RIN ₂₀ OMA, max.	dB/Hz	TBD	
Extinction Ratio, min	dB	TBD	
Transmitter and dispersion penalty (TDP), max	dB	TBD	
Transmitter eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻³ hits per sample			
Receiver (gamma- R)			
Average received power, max.	dBm	TBD	
Rx jitter tracking test, OMA	mW(dBm)	TBD	
Rx jitter tracking test, frequency and pk-pk amplitude	(kHz,UI)	TBD	
Unstressed receiver sensitivity, OMA	mW(dBm)	TBD	
Return loss of receiver, min.	dB	TBD	
SRS eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻⁵ hits per sample			
Notes:			
1 See: IEC 60793-2-50 (reference [10]), Type B1.1 and IEC 60793-2-50 (reference [10]), Type B1.3, and IEC 60793-2-50 (reference [10]), Type B6 Optical fibers - Part 2: Product Specifications.			
2 The signaling rate shall not deviate by more than ±100 ppm from the nominal data rate over all periods equal to 200 000 transmitted bits (~10 max length frames).			
3			

5.4.3.2 256GFC-CWDM4

Table 9 – Single-mode wavelength division multiplex link classes (OS1, OS2)

Single mode wavelength division multiplex link parameters	Unit	256GFC-CWDM4	Note
Nominal signaling rate per wavelength	MBd	115 600	1
Operating distance	m	2 000	2
Transmitter (gamma-T)			
Center wavelength	nm	1264.5-1277.5 1284.5-1297.5 1304.5-1317.5 1324.5-1337.5	
Side mode suppression per wavelength (min)	dB	TBD	
Total launched average power (max)	mW (dBm)	TBD	
Tx average optical power, each wavelength (max)	mW (dBm)	TBD	
Extinction ratio, each wavelength (min)	dB	TBD	
Tx OMA, each wavelength (max)	mW (dBm)	TBD	
Tx OMA, each wavelength, at max TDP (min)	mW (dBm)	TBD	
Tx OMA, each wavelength (min)	mW (dBm)	TBD	
Launch power in OMA minus TDP, each wavelength (min)	mW (dBm)	TBD	
Transmitter and dispersion penalty (TDP), each wavelength (max)	dB	TBD	
Transmitter eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻³ hits per sample			
Channel			
Channel insertion loss (max)	dB	TBD	
Receiver (gamma- R)			
Damage threshold, each wavelength (min)	mW (dBm)	TBD	
Average received power, each wavelength (max)	mW (dBm)	TBD	
Average received power, each wavelength (min)	mW (dBm)	TBD	
Received power (OMA), each wavelength (max)	mW (dBm)	TBD	
Receiver reflectance (max)	dB	TBD	
Rx sensitivity in OMA, each wavelength (max)	mW (dBm)	TBD	
Stressed receiver sensitivity (OMA), each wavelength (max)	mW (dBm)	TBD	
Conditions of stressed receiver sensitivity test			
Vertical eye closure penalty, each wavelength	dB	TBD	
Stressed eye J2 jitter, each wavelength	UI	TBD	
Stressed eye J4 jitter, each wavelength	UI	TBD	
SRS eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻⁵ hits per sample			
Notes:			
1 The signaling rate shall not deviate by more than ±100 ppm from the nominal data rate over all periods equal to 200 000 transmitted bits (~10 max length frames).			
2 Operating distance is given for OS2 cable. See IEC 60793-2-50 (reference [10]), type B1.3, and type B6 Optical fibers - Part 2: Product Specifications. B.1.3 is regular OS2 and B6a is bend insensitive OS2.			

5.4.3.3 256GFC-PSM4

Table 10 – Single-mode parallel fiber link classes (OS1, OS2)

Single mode parallel fiber link parameters	Unit	256GFC-PSM4	Note
Nominal signaling rate per fiber lane	MBd	115 600	1
Operating distance	m	500	2
Transmitter (gamma-T)			
Center wavelength, max.	nm	1325	
Center wavelength, min.	nm	1295	
Side-mode suppression ratio (min)	dB	TBD	
Total average launch power (max)	mW(dBm)	TBD	
Average launch power, each lane (max)	mW(dBm)	TBD	
Average launch power, each lane (min)	mW(dBm)	TBD	
Optical Modulation Amplitude (OMA), each lane (max)	mW(dBm)	TBD	3
Transmitter and dispersion penalty (TDP), each lane (max)	dB	TBD	
Transmit OMA, each lane (min)	mW(dBm)	TBD	
Average launch power of OFF transmitter, each lane (max)	mW(dBm)	TBD	
Extinction ratio (min)	dB	TBD	
Optical return loss tolerance (max)	dB	TBD	
Transmitter reflectance (max)	dB	TBD	
Transmitter eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻³ hits per sample			
Channel			
Channel insertion loss (max)	dB	TBD	
Receiver (gamma-R)			
Damage threshold each lane (min)	mW(dBm)	TBD	
Average receive power, each lane (max)	mW(dBm)	TBD	
Average receive power, each lane (min)	mW(dBm)	TBD	
Receive power, each lane (OMA) (max)	mW(dBm)	TBD	
Receiver reflectance (max)	dB	TBD	
Receiver sensitivity (OMA), each lane (max)	mW(dBm)	TBD	
Stressed receiver sensitivity (OMA), each lane (max)	mW(dBm)	TBD	
Vertical eye closure penalty, each lane	dB	TBD	
Stressed eye J2 jitter, each lane	UI	TBD	
Stressed eye J4 jitter, each lane	UI	TBD	
Stressed eye mask definition {X1,X2,X3,Y1,Y2,Y3}		TBD	
Hit ratio 10 ⁻⁵ hits per sample			
OMA of each aggressor lane	mW(dBm)	TBD	
Notes:			
1 The signaling rate shall not deviate by more than ±100 ppm from the nominal data rate over all periods equal to 200 000 transmitted bits (~10 max length frames).			
2 Operating distance is given for OS2 cable. See IEC 60793-2-50 (reference [10]), type B1.3, and type B6 Optical fibers - Part 2: Product Specifications. B.1.3 is regular OS2 and B6a is bend insensitive OS2.			
3 See FC-MSQS (reference [7]).			
4			

00 5.5 MM data links

01

02 5.5.1 MM general information

03 Table 11 gives the variant names, a general link description, and the gamma compliance point spec-
04 ifications for multi-mode optical fiber links running at 64GFC and 256GFC. The receiver shall operate
05 with a maximum BER of 1.09×10^{-4} when the input power falls within the range given in table 11.
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Table 11 – Multimode link classes

Multimode link parameters	Unit	6400-SW1	25600-SW4	Note
Nominal signaling rate	MBd	28 900	115 600	1
Operating distance (OM3)	m	0.5 - 70	0.5 - 70	2,3,4
Operating distance (OM4)	m	0.5 - 100	0.5 - 100	
Modulation Format		PAM-4	PAM-4	
Transmitter (gamma-T)				
Source type		Laser	Laser	
Center wavelength, min.	nm	840	840	
Center wavelength, max.	nm	860	860	
RMS spectral width, max.	nm	0.6	0.6	
TDECQ (max)	dB	5	5	5
Average launched power, max.	mW(dBm)	2.512 (+4)	2.512 (+4)	6
Average launched power, min.	mW(dBm)	0.251 (-6)	0.251 (-6)	7
OMA _{outer} , max.	mW(dBm)	1.995 (+3)	1.995 (+3)	8
OMA _{outer} , min.	mW(dBm)	0.398 (-4)	0.398 (-4)	8,9
Launch power in OMA _{outer} minus TDECQ (min)	mW(dBm)	0.3162(-5)	0.3162(-5)	
OMA _{outer} extinction ratio (min)	dB	3	3	10
Encircled flux		≥86% at 19 μm ≤30% at 4.5 μm		3
Receiver (gamma- R)				
Damage Threshold, each lane (min)	mW(dBm)	3.162(+5)	3.162(+5)	11
Average received power, max.	mW(dBm)	2.512(+4)	2.512(+4)	
Average received power, min.	mW(dBm)	0.162(-7.9)	0.162(-7.9)	
Receive power (OMA _{outer}) (max), mW (dBm)	mW(dBm)	1.995(+3)	1.995(+3)	
Return Loss of Receiver, min.	dB	12	12	
Stressed receiver sensitivity, OMA _{outer} each lane (max)	mW(dBm)	0.631(-2)	0.631(-2)	
Receiver sensitivity, OMA _{outer} each lane (max)	mW(dBm)	0.1995(-7)	0.1995(-7)	
Stressed test source				
Stressed eye closure (SECQ)	dB	5	5	
OMA of each aggressor	mW(dBm)	NA	1.995(+3)	
Notes:				
1 The signaling rate shall not deviate by more than ±100 ppm from the nominal signaling rate over all periods equal to 200 000 transmitted bits (~10 max length frames).				
2 The operating ranges shown here are based on MM fiber bandwidths given in table 20 of FC-PI-5 (reference [1]) and a 1.5 dB total connector loss. For link budget calculations methodology see FC-MSQS (reference [7]) and FC-MSQS-2 (reference [27]).				
3 Encircled flux specifications in accordance with TIA-492AAAC-A (reference [19]) and IEC 60793-2-10 (reference [9]) or IEEE 802.3 clause 52 (reference [16]).				
4 See ___[] for channel loss (TBD)				
5 Transmitter Dispersion Eye Closure for PAM4 signals, TDECQ, see IEEE 802.3bs (reference [26]). For PI-7, TDECQ, must be computed using a BER of 1.09x10 ⁻⁴ .				
6 Defined by average received power, max				
7 The value is calculated using an infinite extinction ratio at the lowest allowed transmit OMA.				
8 For definitions of Outer Optical Modulation Amplitude, OMA _{outer} , see IEEE 802.3bs (reference [26]).				
9 Even if TDECQ < 1 dbi OMA _{outer} (min) must exceed this value				
10 For definition of OMA _{outer} extinction ratio see [TBD]				
11 The receivers should be able to tolerate, without damage, continuous exposure to an optical input signal having this average power level. The receiver does not have to operate correctly at this received power.				

6 Electrical interface specification - single lane variants

This clause defines the electrical Tx and Rx parameters for the channel between a host ASIC and a transceiver module plugged into a separable connector at the Fibre Channel delta-T/delta-R compliance points. The existence of a compliance point is determined by the existence of a connector at that point in a TxRx connection. Annex C provides the channel electrical characteristics. Significant material from OIF CEI-28G-VSR (reference [22]) was utilized in developing this clause.

6.1 General electrical characteristics

Each conforming electrical FC device shall be compatible with this serial electrical interface to allow interoperability within an FC environment. Fibre Channel 64GFC and 256GFC links use the 256B/257B transmission code; see FC-FS-4 (reference [23]) and see FC-FS-5 (reference [25]). This code includes Forward Error Correction which is required to achieve the link BER objective. Prior to error correction, Fibre Channel 64GFC TxRx connections shall not exceed a BER of 1.09×10^{-4} under any compliant conditions. The parameters in this clause support meeting that requirement. At this level of BER performance, it is expected that the BER after error correction will be undetectably low.

TxRx connections operating at these maximum distances may require some form of equalization to enable the signal requirements to be met. Greater distances may be obtained by specifically engineering a TxRx connection based on knowledge of the technology characteristics and the conditions under which the TxRx connection is installed and operated. However, such distance extensions are outside the scope of this standard. The general electrical characteristics are described in table 12.

Table 12 – General electrical characteristics

	Units	6400-DF-EA-S	25600-DF-EA-S
Data rate (note 1)	MB/s	6 400	256 000
Nominal symbol rate	MBd	28 900	28 900
Tolerance	ppm	±100	±100
Differential Impedance	Ω (nom)	100	100
Notes:			
1 The data rate may be verified by determining the time to transmit at least 200 000 transmission bits (10 max length FC frames).			

6.2 Compliance test point definitions

6.2.1 Test method

The interoperability points are generally defined for Fibre Channel systems as being immediately after the mated connector. For the delta points this is not an easy measurement point, particularly at high frequencies, as test probes cannot be applied to these points without affecting the signals being measured, and de-embedding the effects of test fixtures is difficult. For delta point measurements reference test points are defined with a set of defined test boards for measurement consistency. The delta point specifications in FC-PI-6 are to be interpreted as being at the RF connector outputs and inputs of the reference compliance boards.

In order to provide test results that are reproducible and easily measured, this document defines two test boards that have RF connector interfaces for easy connection to test equipment. One is designed for insertion into a host, and one for inserting modules. The reference test boards' objectives are:

- Satisfy the need for interoperability at the electrical level.

- Allow for independent validation of host and module.
- The PCB traces are targeted at $100\ \Omega$ differential impedance with nominal 7% differential coupling.

Testing compliance to specifications in a high-speed system is delicate and requires thorough consideration. Using common test boards that allow predictable, repeatable, and consistent results among vendors will help to ensure consistency and true compliance in the testing.

The reference test boards provide a set of overlapping measurements for module and host validation to ensure system interoperability.

6.2.2 Host test points

Host system transmitter and receiver compliance are defined by tests in which a Host Compliance Board is inserted, as shown in figure 5, in place of the module. The test points are B and C.

Host compliance points are defined as the following:

- B: host output at the output of the Host Compliance Board. Electrical output and host return loss specifications shall be met at this point.
- C: host input at the input of the Host Compliance Board. Host return loss specifications shall be met at this point.

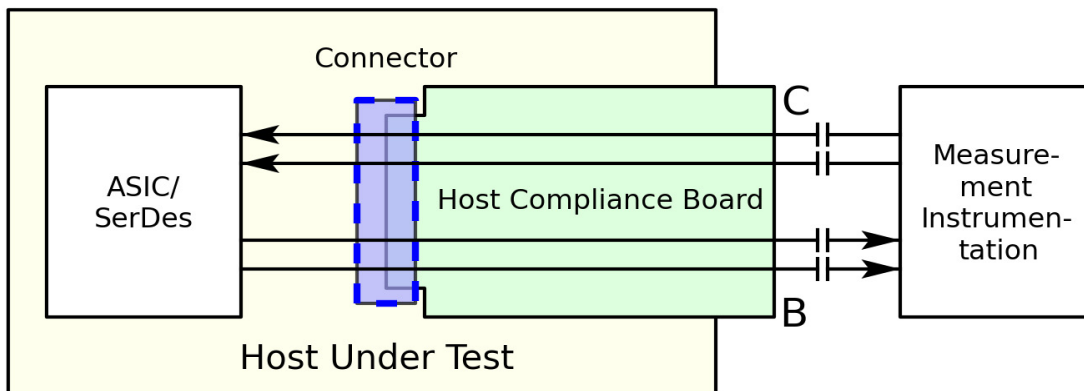


Figure 5 – Host Compliance Board

6.2.3 Module test points

Module transmitter and receiver compliance are defined by tests in which the module is inserted into the Module Compliance Board as shown in figure 6.

Module test points are defined as the following:

- B': Module electrical input at the input of the Module Compliance Board. Module return loss specifications shall be met at this point.
- C': Module electrical output at the output of the Module Compliance Board. Module output and module return loss specifications shall be met at this point.

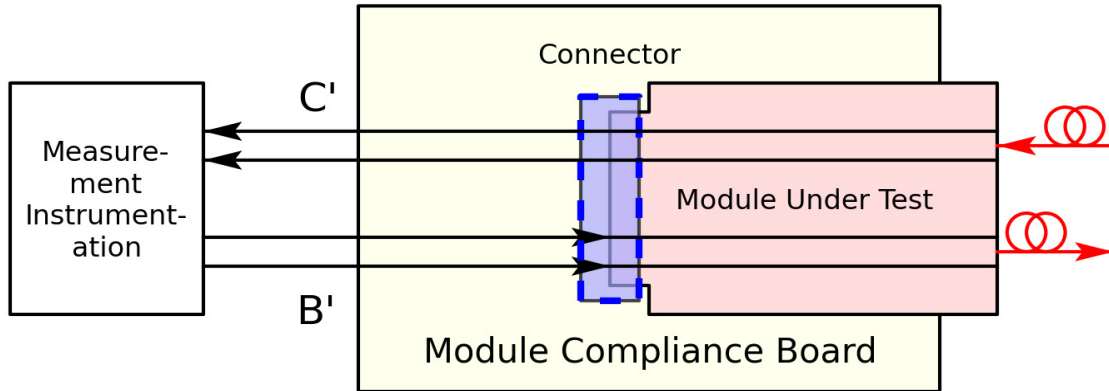


Figure 6 – Module Compliance Board

6.2.4 Host input calibration point

The host receiver input tolerance signal is calibrated through the Host Compliance Board at the output of the Module Compliance Board as shown in figure 7. The opposite data path is excited with an asynchronous test source with PRBS31 or 64GFC IDLE. See table 14 for electrical characteristics. The host input calibration point is at C'' with specifications for input signals being calibrated at C''. Note that the point C'' has additional trace loss beyond the edge connector pins.

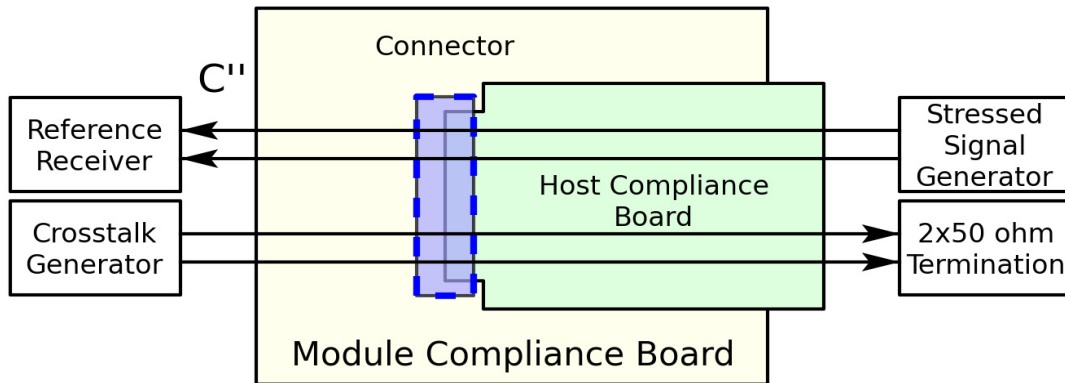


Figure 7 – Host input calibration point C''

6.2.5 Module input calibration point

The module electrical input tolerance signal is calibrated through the Module Compliance Board at the output of the Host Compliance Board as shown in figure 8. The opposite data path is excited with an asynchronous test source with PRBS31 or 64GFC IDLE. See table 14 for electrical characteristics. The module input calibration point is at B'' with specifications for input signals being calibrated at B''. Note that point B'' has additional trace loss beyond the module pins.

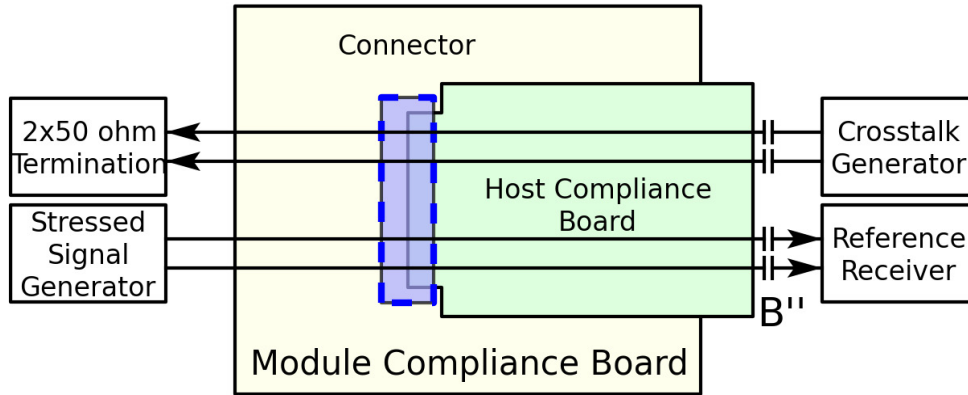


Figure 8 – Module input calibration point B''

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6.3 Transmitted signal characteristics

This subclause defines the interoperability requirements of the transmitted signal at the driver end of a TxRx connection. Details for the measurement process are specified in FC-MSQS-2 (reference [27]), as adapted from OIF (reference [18]).

Hosts and modules shall meet the appropriate specifications defined in table 13.

Table 13 – Transmitter compliance requirements

Parameter	Host output		Module electrical output		Units	Notes
	Min	Max	Min	Max		
Compliance point	B (figure 5)		C' (figure 6)			note 1
Differential voltage pk-pk	-	TBD		TBD	mV	
Common mode noise rms	-	TBD		TBD	mv	
Differential termination resistance mismatch	-	TBD		TBD	%	note 2
Differential return loss SDD22	-	equation 1 figure 10	-	equation 1 figure 10	dB	note 3
Common mode to differential conversion SDC22	-	equation 3 figure 12	-	equation 3 figure 12	dB	note 4
Differential to common mode conversion SCD22						
Common mode return loss SCC22	-	TBD		TBD	dB	note 5
Source transition time 20%-80%		TBD		TBD	ps	
Common mode voltage		TBD		TBD	V	note 6
Vertical eye closure		TBD		TBD	dB	note 7
Eye width at 10^{-6} probability EW6		TBD		TBD	UI	
Eye height at 10^{-6} probability EH6		TBD		TBD	mV	
Eye Linearity		TBD		TBD		note 8
Crosstalk parameters						
Signal calibration point	C" (figure 7)		B" (figure 8)			note 1, note 9
Signal application point	C (figure 5)		B' (figure 6)			
Crosstalk amplitude differential voltage pk-pk	TBD		TBD		mV	
Crosstalk transition time 20%-80%	TBD		TBD		ps	note 10
Notes:						
1 See compliance test point definitions in subclause 6.2						
2 At 1 MHz						
3 See subclause 6.6.1 for differential return loss SDD22						
4 See subclause 6.6.2 for common mode to differential conversion SCD22 and differential to common mode conversion SDC22						
5 From 250 MHz to 30 GHz						
6 Referred to host ground						
7 Open eye is generated through the use of a Continuous Time Linear Equalizer (CTLE). See FC-MSQS-2 (reference [27]) for test configurations and test methods. The module may need equalization to achieve the required eye opening						
8 Eye linearity = $(\max(AV_{\text{upp}}, AV_{\text{mid}}, AV_{\text{low}}) / \min(AV_{\text{upp}}, AV_{\text{mid}}, AV_{\text{low}}))$.						
9 Host crosstalk calibration is specified by Figure 3.1 (diagram on the right) and clause 3.2.3; module crosstalk calibration is specified by Figure 3.3 (diagram on the right) and clause 3.2.3 of FC-MSQS-2 (reference [27])						
10 Crosstalk transition times are measured at the input of the compliance test board						

6.4 Receive signal characteristics

This subclause defines the interoperability requirements of the delivered signal at the receive device end of a TxRx connection. Details for the measurement process are specified in FC-MSQS-2 (reference [27]), as adapted from OIF (reference [17]).

Hosts and modules shall meet the appropriate specifications defined in table 14.

Table 14 – Receiver compliance requirements

Parameter	Host input		Module electrical input		Units	Notes
	Min	Max	Min	Max		
Return loss, mode conversion, and common mode voltage requirements						
Compliance point	C (figure 5)		B' (figure 6)			note 1
Differential termination resistance mismatch	-	TBD		TBD	%	
Differential return loss SDD11	-	equation 1 figure 10	-	equation 1 figure 10	dB	note 2
Common mode to differential conversion SDC11	-	equation 2 figure 11	-	equation 2 figure 11	dB	note 3
Differential mode to common conversion SCD11						
Common mode voltage		TBD		TBD	V	note 4
Crosstalk signal requirements						
Signal calibration point	B" (figure 8)		C" (figure 7)			note 1, note 5
Signal application point	B (figure 5)		C' (figure 6)			
Crosstalk amplitude differential voltage pk-pk	TBD		TBD		mV	
Crosstalk source transition time 20%-80%	TBD		TBD		ps	note 6
Stressed receiver test requirements						
Signal calibration point	C" (figure 7)		B" (figure 8)			note 1
Signal application point	C (figure 5)		B' (figure 6)			
Random jitter, peak-to-peak, 10 ⁻⁶ BER	-	TBD		TBD	UI	note 7, note 8
Eye width at 10 ⁻⁶ probability EW6		TBD		TBD	UI	
Eye height at 10 ⁻⁶ probability EH6		TBD		TBD	mV	
Notes:						
1 See compliance test point definitions in subclause 6.2						
2 See subclause 6.6.1 for differential return loss SDD22						
3 See subclause 6.6.2 for common mode to differential conversion SCD22 and differential to common mode conversion SDC22						
4 Referred to host ground. Common mode voltage is generated by the host.						
5 During the module electrical input test, the crosstalk signal is generated by the module from an incoming optical signal. For purposes of calibrating the module stress signal, a worst case crosstalk signal is required to be produced by an electronic signal generator. See FC-MSQS-2 (reference [27]).						
6 Crosstalk transition times are measured at the input of the compliance test board						
7 Uncorrelated bounded jitter is added to meet the EW6 requirement at 10 ⁻⁶ . See FC-MSQS-2 (reference [27]) for test configurations and test methods.						
8 Host crosstalk calibration is specified by Figure 3.2 (diagram on the right) and clause 3.2.3; module crosstalk calibration is specified by Figure 3.4 (diagram on the right) and clause 3.2.3 of FC-MSQS-2 (reference [27])						

6.5 Receive jitter tracking compliance

The sinusoidal jitter tracking test measures the ability of the receiver to track low frequency jitter. This test is done without other added jitter impairments to understand the ability of the device under test to track low frequency jitter.

Table 15 – Minimum jitter tolerance

Parameter	Low	High
Jitter frequency	TBD	TBD
Jitter amplitude, pk-pk	TBD	TBD

The following figure shows the jitter tracking template:

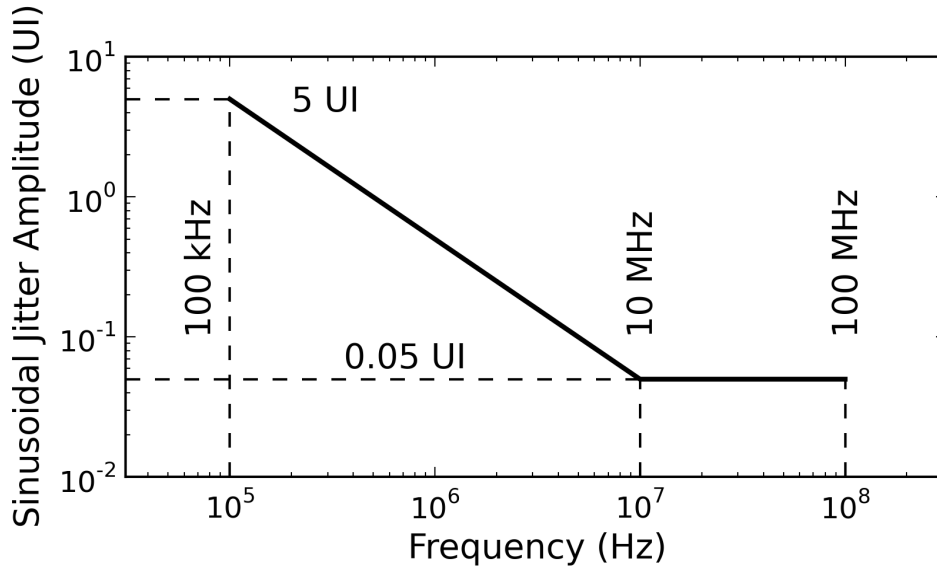


Figure 9 – Receiver jitter tracking template

For additional testing details see MSQS-2 (reference reference [27]).

00	6.6 Differential return loss and mode conversion requirements	00
01		01
02	6.6.1 Differential return loss	02
03	When measured at the respective test point, return loss shall not exceed the limits given in equation	03
04	1 as illustrated in figure 10.	04
05		05
06		06
07		07
08		08
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10		10
11		11
12		12
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25	Figure 10 – SDD11 and SDD22 for all compliance points	25
26		26
27		27
28	Return loss equation at the appropriate test points:	28
29	$SDD11, SDD22 \text{ (dB)} < \{ TBD \}$	(1) 29
30		30
31	6.6.2 Common to differential mode and differential to common mode conversion	31
32		32
33	The common to differential mode and differential to common mode conversion specifications are intended to limit the amount of unwanted signal energy that is allowed to be generated due to conversion of common mode voltage to differential mode voltage or vice versa.	33
34		34
35		35
36	When measured at the respective test point, common to differential mode or differential to common mode conversion SDC11 and SCD11 shall not exceed the limits given in equation 2 as shown in figure 11.	36
37		37
38		38
39		39
40		40
41	$SDC11, SCD11 \text{ (dB)} < \{ TBD \}$	(2) 41
42		42
43	When measured at the respective test point, common to differential mode or differential to common mode conversion SDC22 and SCD22 shall not exceed the limits given in equation 3 as shown in figure 12.	43
44		44
45		45
46	$SDC22, SCD22 \text{ (dB)} < \{ TBD \}$	(3) 46
47		47
48		48
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Figure 11 – SDC11 and SCD11 for receiver input

Figure 12 – SDC22 and SCD22 for transmitter output